

Codes and Cryptography

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PART VIII

Outline

- 1 **BCH Codes**
- 2 Concatenation
- 3 Interleaving
- 4 Real Systems

Why Looking for More Cyclic Codes?

Reed-Solomon codes can be seen as MDS cyclic codes. But over a fixed field \mathbb{F}_q , MDS codes of arbitrary length $n > q + 2$, are conjectured to exist only when $k = 1$ (the repetition code) or $k \geq n - 1$ (that is $d = 1$ or 2).

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E.g., BCH codes.

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If $e = 1$, then $n \mid q - 1$, $\deg g = m$, $d = m + 1$ and \mathcal{C} is MDS.

Proof of the lower bound $d \geq m + 1$

Assume that there is a codeword of weight m (or less)

$$w = \lambda_1 X^{i_1} + \dots + \lambda_m X^{i_m}, \text{ where } i_1 < \dots < i_m$$

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This means that $(\lambda_1, \dots, \lambda_m)$ is a solution to

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The same proof works for any set of roots of the form $\beta^a, \beta^{a+b}, \dots, \beta^{a+bm}$ for $\gcd(b, n) = 1$.

Example: $[17, 9, 5]_2$ BCH Code

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Conjugate classes: $\{1\}$ $\{\beta, \beta^2, \beta^4, \beta^8, \beta^{16}, \beta^{15}, \beta^{13}, \beta^9\}$

$\{\beta^3, \beta^6, \beta^{12}, \beta^7, \beta^{14}, \beta^{11}, \beta^5, \beta^{10}\}$

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Taking $m = 2$: (The design distance is $d^* = m + 1 = 3$)

$$g = (X - \beta)(X - \beta^2)(X - \beta^4)(X - \beta^8)(X - \beta^{16})(X - \beta^{15})(X - \beta^{13})(X - \beta^9) =$$

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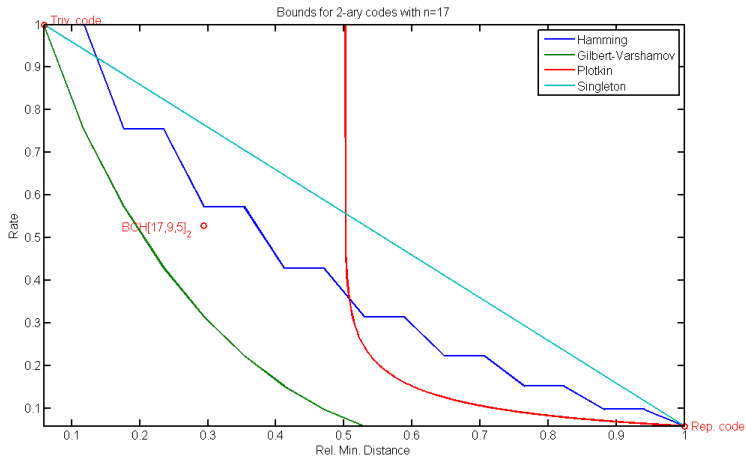
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\mathcal{C} is a $[17, 9, 5]_2$ BCH code, with an optimal dimension.

The sphere packing bound reads here $k \leq \left\lfloor 17 - \log_2(1 + 17 + \binom{17}{2}) \right\rfloor = 9$

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- \mathcal{C}' is an MDS code,
- encoding in \mathcal{C}' can be computed by division by g ,
- error correction in \mathcal{C}' works the same way as in the original cyclic Reed-Solomon code \mathcal{C} .

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Because symbols are stored into a two dimensional structure, and there are new adjacency relations.

This makes errors accumulate in a single codeword, typically exceeding the error correction capability of the code.

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Build a complex code from two simpler ones by composition of the encoding functions. Then apply the two layers of error correction.

The resulting encoding function is simpler than the encoding function of a directly designed large code.

Correcting a large number of errors is also more efficient.

Adds protection against some specific error patterns.

Concatenation of Linear Codes (I)

Let $\mathcal{C}_1 \subset (\mathbb{F}_{q_1})^{n_1}$ be a $[n_1, k_1, d_1]_{q_1}$ -linear code, called the **inner code**, and let $\mathcal{C}_2 \subset (\mathbb{F}_{q_2})^{n_2}$ be a $[n_2, k_2, d_2]_{q_2}$ -linear code, called the **outer code**.

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We assume the existence of a q -linear bijection $\psi : \mathbb{F}_{q_2}^{n_2} \rightarrow \mathbb{F}_{q_1}^{k_1}$

Thus, $q_1 = q^{e_1}$, $q_2 = q^{e_2}$ for some q , e_1, e_2 and $e_1 k_1 = e_2 n_2$.

We also use natural q -linear bijections $\mathbb{F}_{q^{e_1}} \cong \mathbb{F}_q^{e_1}$ and

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Observe that the encoding map for \mathcal{C} is the composition:

$$\mathbb{F}_q^{e_2 k_2} \cong (\mathbb{F}_{q^{e_2}})^{k_2} \xrightarrow{\text{Enc}_2} (\mathbb{F}_{q^{e_2}})^{n_2} \cong (\mathbb{F}_{q^{e_1}})^{k_1} \xrightarrow{\text{Enc}_1} (\mathbb{F}_{q^{e_1}})^{n_1} \cong \mathbb{F}_q^{e_1 n_1}$$

Concatenation of Linear Codes (II)

As a particular case, $e_1 = 1$, $e_2 = k'_1$ and C_1 consists of n_2 copies of a $[n'_1, k'_1, d'_1]$ code C'_1 (Each symbol in the outer codeword is encoded separately with C'_1)

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$s_{11} \cdots s_{1k'_1} s_{21} \cdots s_{2k'_1} \cdots s_{k_2 1} \cdots s_{k_2 k'_1}$

Source sequence in $\mathbb{F}_q^{k'_1 k_2}$

Concatenation of Linear Codes (II)

As a particular case, $e_1 = 1$, $e_2 = k_1'$ and C_1 consists of n_2 copies of a $[n_1', k_1', d_1']$ code C_1' (Each symbol in the outer codeword is encoded separately with C_1')

Then, C_1 is $[n_1' n_2, k_1' k_2, d_1']_q$ and $d \geq d_1' d_2$.

$$\begin{array}{ccccccc}
 s_{11} & \cdots & s_{1k_1'} & s_{21} & \cdots & s_{2k_1'} & \cdots & s_{k_2 1} & \cdots & s_{k_2 k_1'} \\
 \mathbf{s}_1 & & & \mathbf{s}_2 & & & \cdots & & & \mathbf{s}_{k_2}
 \end{array}$$

Source sequence in $(\mathbb{F}_q^{k_1'})^{k_2} \cong (\mathbb{F}_q^{e_2})^{k_2}$

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 s_{11} & \cdots & s_{1k'_1} & s_{21} & \cdots & s_{2k'_1} & \cdots & s_{k_2 1} & \cdots & s_{k_2 k'_1} \\
 \mathbf{s}_1 & & \mathbf{s}_2 & & \cdots & & \mathbf{s}_{k_2} & & & \\
 \mathbf{x}_1 & & \mathbf{x}_2 & & \cdots & & \mathbf{x}_{k_2} & & \cdots & \mathbf{x}_{n_2}
 \end{array}$$

Outer codeword in $C_2 \subset (\mathbb{F}_{q^{e_2}})^{n_2} \cong (\mathbb{F}_q^{k'_1})^{n_2}$

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$$\begin{array}{ccccccc}
 s_{11} & \cdots & s_{1k_1'} & s_{21} & \cdots & s_{2k_1'} & \cdots & s_{k_21} & \cdots & s_{k_2k_1'} \\
 \mathbf{s}_1 & & \mathbf{s}_2 & & \cdots & \mathbf{s}_{k_2} & & & & \\
 \mathbf{x}_1 & & \mathbf{x}_2 & & \cdots & \mathbf{x}_{k_2} & & \cdots & & \mathbf{x}_{n_2} \\
 \mathbf{y}_1 & & \mathbf{y}_2 & & \cdots & \mathbf{y}_{k_2} & & \cdots & & \mathbf{y}_{n_2}
 \end{array}$$

Inner codeword sequence in $(C_1')^{n_2} \subset (\mathbb{F}_q^{n_1'})^{n_2}$

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 \mathbf{s}_1 & \mathbf{s}_2 & \cdots & \mathbf{s}_{k_2} & & & \\
 \mathbf{x}_1 & \mathbf{x}_2 & \cdots & \mathbf{x}_{k_2} & \cdots & \mathbf{x}_{n_2} & \\
 \mathbf{y}_1 & \mathbf{y}_2 & \cdots & \mathbf{y}_{k_2} & \cdots & \mathbf{y}_{n_2} & \\
 y_{11} \cdots y_{1k'_1} \cdots y_{1n'_1} & y_{21} \cdots y_{2k'_1} \cdots y_{2n'_1} & \cdots & y_{k_2 1} \cdots y_{k_2 k'_1} \cdots y_{k_2 n'_1} & \cdots & y_{n_2 1} \cdots y_{n_2 n'_1} &
 \end{array}$$

Final encoded sequence in $\mathbb{F}_q^{n'_1 n_2}$

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 \mathbf{y}_1 & \mathbf{y}_2 & \cdots & \mathbf{y}_{k_2} & \cdots & \mathbf{y}_{n_2} & \\
 y_{11} \cdots y_{1k'_1} \cdots y_{1n'_1} & y_{21} \cdots y_{2k'_1} \cdots y_{2n'_1} & \cdots & y_{k_2 1} \cdots y_{k_2 k'_1} \cdots y_{k_2 n'_1} & \cdots & y_{n_2 1} \cdots y_{n_2 n'_1} &
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 \mathbf{x}_1 & \mathbf{x}_2 & \cdots & \mathbf{x}_{k_2} & \cdots & \mathbf{x}_{n_2} & \\
 \mathbf{y}_1 & \mathbf{y}_2 & \cdots & \mathbf{y}_{k_2} & \cdots & \mathbf{y}_{n_2} & \\
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This is just looking to a $[n, k, d]_q$ code as a $[en, ek, ed]_q$ code, which has the same rate, the same random error correction capability, but it can correct up to $\lfloor \frac{d_2 - 1}{4} \rfloor$ bursts of length e .

Example: Inner $15 \times [17, 9, 5]_2$, Outer $[15, 11, 5]_{2^9}$

Inner code: 15 copies of the previous example of $[17, 9, 5]_2$
BCH code

- it corrects up to 2 errors (each copy)
- it detects up to 4 errors (each copy)

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The concatenated code is a $[255, 99, d]_2$ code with $d \geq 25$.

- it corrects at least any 12 errors
- it can handle more errors if they follow some pattern...

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Decoding Strategy 1: Use the inner code to correct up to 2 errors per inner codeword. If error correction is impossible, mark the inner codeword as erasure.

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Bad inner decoding can only happen from 4 errors. All inner codeword with 2 or 3 errors are marked as erasures.

Bad outer decoding typically happens when there are more than 1 error in more than 4 inner codewords.

Concatenation of Linear Codes (III)

Another approach is using $e_1 = e_2 = 1$ and then $n_2 = k_1$. The encoding sequence is just

$$\mathbb{F}_q^{k_2} \xrightarrow{Enc_2} (\mathbb{F}_q)^{n_2} = (\mathbb{F}_q)^{k_1} \xrightarrow{Enc_1} (\mathbb{F}_q)^{n_1}$$

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For instance, we can use a $[28, 24, 5]_{2^8}$ shortened Reed-Solomon code as outer code and a $[32, 28, 5]_{2^8}$ shortened Reed-Solomon code as inner code.

Each layer adds four redundancy symbols (1024 bits) and can correct up to 2 errors or 4 erasures.

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...but the benefits come when concatenation is combined with interleaving...

Outline

- 1 BCH Codes
- 2 Concatenation
- 3 Interleaving**
- 4 Real Systems

Interleaving

Interleaving is just shuffling the symbols in the codewords in order to break the adjacency relations.

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For instance, the simplest way to interleave m codewords $\mathbf{x}_1, \dots, \mathbf{x}_m \in \mathbf{F}_q^n$ is writing them as rows in a matrix and reading the matrix columnwise.

$$\begin{pmatrix} x_{11} & x_{12} & \dots & x_{1n} \\ x_{21} & x_{22} & \dots & x_{2n} \\ \vdots & \vdots & & \vdots \\ x_{m1} & x_{m2} & \dots & x_{mn} \end{pmatrix} \longrightarrow (x_{11}, x_{21}, \dots, x_{m1}, \dots, x_{mn})$$

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Then, an error burst of length $\leq r$ can only cause up to $\lceil \frac{r}{m} \rceil$ errors in the same codeword.

Cross Interleaving

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Starting from a sequence of codewords $\mathbf{x}_1, \mathbf{x}_2, \dots \in \mathcal{C} \subset \mathbb{F}_q^n$, or the corresponding sequence of symbols in \mathbb{F}_q

$$\dots x_{1,1} x_{1,2} \dots x_{1,n} x_{2,1} x_{2,2} \dots x_{2,n} x_{3,1} x_{3,2} \dots x_{3,n} \dots$$

we construct the interleaved sequence of symbols in \mathbb{F}_q

$$\dots x_{n,1} x_{n-1,2} \dots x_{1,n} x_{n+1,1} x_{n,2} \dots x_{2,n} x_{n+2,1} x_{n+1,2} \dots x_{3,n} \dots$$

which can be seen as applying delays $0, +1, +2, \dots, +(n-1)$ to the symbols in the codeword.

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which can be seen as applying delays $0, +1, +2, \dots, +(n-1)$ to the symbols in the codeword.

An error burst of length $n+1$ is deinterleaved into single errors affecting $n+1$ codewords, which will be easily corrected.

Example of Cross Interleaving

For example, consider $n = 4$, the codeword \mathbf{x}_i is represented as (a_i, b_i, c_i, d_i) , and undefined symbols are set to 0 (this only applies to the start and end of the stream).

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$$\begin{array}{l}
 \dots 00a_1 a_2 a_3 a_4 a_5 \dots \\
 \dots 00b_1 b_2 b_3 b_4 b_5 \dots \\
 \dots 00c_1 c_2 c_3 c_4 c_5 \dots \\
 \dots 00d_1 d_2 d_3 d_4 d_5 \dots
 \end{array}
 \rightarrow a_1 000 a_2 b_1 00 a_3 b_2 c_1 0 a_4 b_3 c_2 d_1 a_5 b_4 c_3 d_2 \dots$$

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 \dots 00c_1 c_2 c_3 c_4 c_5 \dots \\
 \dots 00d_1 d_2 d_3 d_4 d_5 \dots
 \end{array}
 \rightarrow
 a_1 000 a_2 b_1 00 a_3 b_2 c_1 0 a_4 b_3 c_2 d_1 a_5 b_4 c_3 d_2 \dots$$

An error burst of length 5 is transformed into single errors in 5 consecutive codewords.

Concatenation and Interleaving (I)

In practice, interleaving and concatenation are combined to achieve resilience to long error bursts.

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Consider now cross interleaving after encoding with a $[6, 4, 3]$ inner code.

$$\begin{array}{l}
 a_1 a_2 a_3 a_4 a_5 a_6 a_7 a_8 a_9 \dots \\
 b_1 b_2 b_3 b_4 b_5 b_6 b_7 b_8 b_9 \dots \\
 c_1 c_2 c_3 c_4 c_5 c_6 c_7 c_8 c_9 \dots \\
 d_1 d_2 d_3 d_4 d_5 d_6 d_7 d_8 d_9 \dots \\
 e_1 e_2 e_3 e_4 e_5 e_6 e_7 e_8 e_9 \dots \\
 f_1 f_2 f_3 f_4 f_5 f_6 f_7 f_8 f_9 \dots
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 b_1 & b_2 & b_3 & b_4 & b_5 & b_6 & b_7 & b_8 & b_9 & \dots \\
 c_1 & c_2 & c_3 & c_4 & c_5 & c_6 & c_7 & c_8 & c_9 & \dots \\
 d_1 & d_2 & d_3 & d_4 & d_5 & d_6 & d_7 & d_8 & d_9 & \dots \\
 e_1 & e_2 & e_3 & e_4 & e_5 & e_6 & e_7 & e_8 & e_9 & \dots \\
 f_1 & f_2 & f_3 & f_4 & f_5 & f_6 & f_7 & f_8 & f_9 & \dots \\
 1 & 2 & 2 & 2 & 3 & 2 & 1 & 0 & 0 & \dots
 \end{array}$$

An error burst of length 13 produces up to three errors in the same inner codeword.

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$b_1 b_2 b_3 b_4 b_5 b_6 b_7 b_8 b_9 \dots$		$b_1 b_2 b_3 b_4 b_5 b_6 b_7 b_8 b_9 \dots$
$c_1 c_2 c_3 c_4 c_5 c_6 c_7 c_8 c_9 \dots$	→	$c_1 c_2 c_3 c_4 c_5 c_6 c_7 c_8 c_9 \dots$
$d_1 d_2 d_3 d_4 d_5 d_6 d_7 d_8 d_9 \dots$		$d_1 d_2 d_3 d_4 d_5 d_6 d_7 d_8 d_9 \dots$
$e_1 e_2 e_3 e_4 e_5 e_6 e_7 e_8 e_9 \dots$		
$f_1 f_2 f_3 f_4 f_5 f_6 f_7 f_8 f_9 \dots$		
$1 \ 2 \ 2 \ 2 \ 3 \ 2 \ 1 \ 0 \ 0 \ \dots$		

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Concatenation and Interleaving (II)

With another layer of interleaving (e.g., double delay cross interleaving) after encoding with an outer code of minimum distance 4 (e.g., the repetition code $[4, 1, 4]$) all errors and erasures are successfully corrected!

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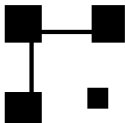
...	a_{-4}	a_{-3}	a_{-2}	a_{-1}	a_0	a_1	a_2	a_3	a_4	a_5	a_6	...	
...	b_{-2}	b_{-1}	b_0	b_1	b_2	b_3	b_4	b_5	b_6	b_7	b_8	...	
...	c_0	c_1	c_2	c_3	c_4	c_5	c_6	c_7	c_8	c_9	c_{10}	...	
...	d_2	d_3	d_4	d_5	d_6	d_7	d_8	d_9	c_{10}	c_{11}	c_{12}	...	
...	1	1	2	1	3	1	3	1	2	0	1	...	(erasures)
...	0	0	0	1	0	1	0	1	0	1	0	...	(errors)

Outline

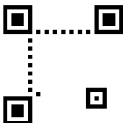
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QR Barcodes Interleaving (I)

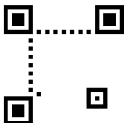
Reserved Areas



Alignment Marks



No Version Info Added



QR Unmasked



with Format Info



Mask



QR Code (w/o border)



QR Code (Final)



Interleaving Map



QR Code version number: 3

QR Code size (w/o border): 29 × 29

Available space: 70 bytes

Error correction level: H

Using field $\mathbb{F}_2[X]/(X^8 + X^4 + X^3 + X^2 + 1)$ for data encoding

Using 2 blocks of shortened RS code $[35, 13, 23]_{256}$

Generating polynomial coeffs. (in hexadecimal)

01 59 B3 83 B0 B6 F4 13 BD 45 28 1C 89 1D 7B 43
FD 56 DA E6 1A 91 F5

Total raw capacity: 26 bytes

Total parity bytes: 44

Data preencoded in 'binary' format

Encoded blocks (in hexadecimal):

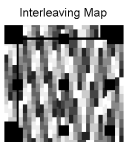
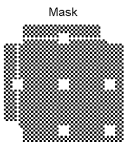
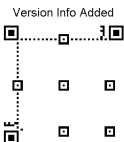
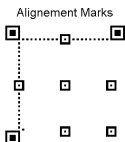
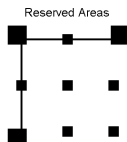
40 E4 86 56 C6 C6 F2 05 76 F7 26 C6 42 D8 BB 91
43 DE F7 47 6F AE 88 42 42 A4 4B A5 5C E5 25 81
3C 55 08

12 12 10 EC 11 EC 11 EC 11 EC 11 EC 11 F6 7B 07
19 07 03 1E 79 F5 84 9C 12 31 DB 03 D3 FA 66 C8
A4 86 AE

Interleaved blocks (in hexadecimal):

40 12 E4 12 86 10 56 EC C6 11 C6 EC F2 11 05 EC
76 11 F7 EC 26 11 C6 EC 42 11 D8 F6 BB 7B 91 07
43 19 DE 07 F7 03 47 1E 6F 79 AE F5 88 84 42 9C
42 12 A4 31 4B DB A5 03 5C D3 E5 FA 25 66 81 C8
3C A4 55 86 08 AE

QR Barcodes Interleaving (II)



QR Code version number: 10
 QR Code size (w/o border): 57×57
 Available space: 346 bytes
 Error correction level: H
 Using field $\mathbb{F}_2[X]/(X^8 + X^4 + X^3 + X^2 + 1)$ for data encoding
 Using 6 blocks of shortened RS code $[43, 15, 29]_{256}$
 and 2 blocks of $[44, 16, 29]_{256}$
 Generating polynomial coeffs. (in hexadecimal)
 01 FC 09 1C 0D 12 FB D0 96 67 AE 64 29 A7 0C F7
 38 75 77 E9 7F B5 64 79 93 B0 4A 3A C5
 Total raw capacity: 122 bytes
 Total parity bytes: 224
 Data preencoded in 'binary' format
 Encoded blocks (in hexadecimal):
 40 00 E4 86 56 C6 C6 F2 05 76 F7 26 C6 42...
 12 10 EC 11 EC 11 EC 11 EC 11 EC 11 EC 11...
 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC...
 EC 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC 11...
 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC...
 EC 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC 11...
 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC...
 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC 11 EC...
 Interleaved blocks (in hexadecimal):
 40 12 11 EC 11 EC 11 11 00 10 EC 11 EC 11 EC EC
 E4 EC 11 EC 11 EC 11 11 86 11 EC 11 EC 11 EC EC
 56 EC 11 EC 11 EC 11 11 C6 11 EC 11 EC 11 EC EC
 C6 EC 11 EC 11 EC 11 11 F2 11 EC 11 EC 11 EC EC
 05 EC 11 EC 11 EC 11 11 76 11 EC...

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Extra error correction can be performed if the source is redundant (and not compressed). E.g., interpolation of missing samples in digital audio produces reasonable results.

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There can be physical limitations that forces structuring the encoded data in a particular way.

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There can be physical limitations that forces structuring the encoded data in a particular way.

The information stored in a Compact Disk can be modelled by a binary string in which ones are always separated by a number of zeros between 2 and 10. Thus, codewords must be reencoded with a special code (EFM or Eight-To-Fourteen Modulation) to fulfil these requirements.

Eight-To-Fourteen Modulation

It is a nonlinear binary code of length 14 and size 257 (not intended for error correction).

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The code is

```
00000000100000 00000000100001 00000000100010 00000000100100 00000001000000 00000001000001  
00000001000010 00000001000100 00000001001000 00000001001001 00000010000000 00000010000001  
00000010000010 00000010000100 00000010001000 00000010001001 ... 10010010010010
```

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Only 256 codewords are used (to map every element in \mathbb{F}_{256} to the code).

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Codewords are separated by three extra binary digits (to avoid breaking the rules about the number of separating zeros).

Codes and Cryptography

Jorge L. Villar

MAMME, Fall 2015

END OF PART VIII